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ZN414Z & ZN416E

AM RADIO RECEIVERS

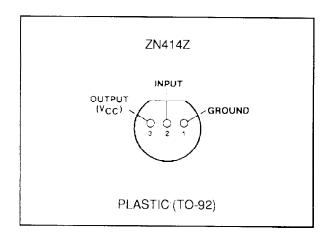
The ZN414Z is a 10 transistor tuned radio frequency (TRF) circuit packaged in a 3-pin TO-92 plastic package for simplicity and space economy .

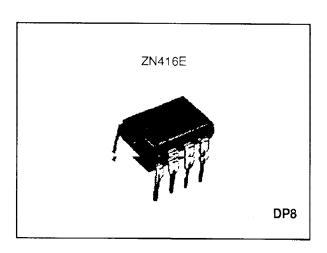
The circuit provides a complete R.F. amplifier, detector and AGC circuit which requires only six external components to give a high quality A.M. tuner. Effective AGC action is available and is simply adjusted by selecting one external resistor value. Excellent audio quality can be achieved, and current consumption is extremely low. No setting-up or alignment is required and the circuit is completely stable in use.

The ZN416E is a buffered output version of the ZN414Z giving typically 120mV (r.m.s.) output into a 64Ω load.

FEATURES

- Single cell operation (1.1 to 1.6 volt operating range)
- Low current consumption
- 150kHz to 3MHz frequency range (i.e. full coverage of medium and long wavebands)
- Easy to assemble, no alignment necessary
- Simple and effective AGC action
- Will drive crystal earphone direct (ZN414Z)
- Will drive headphone direct (ZN416E)
- Excellent audio quality
- Typical power gain of 72dB (ZN414Z)
- Minimum of external components required





DEVICE SPECIFICATIONS $(T_{amb} = 25^{\circ}C, V_{cc} = 1.4V. Parameters apply to all types unless otherwise specified)$

| Parameter | | Min. | Тур. | Max. | Units |
|--|--------|------|----------|------|-----------|
| Supply voltage, V _{cc} | | 1.1 | 1.3 | 1.6 | Volts |
| Supply current, I $_{ m s}$ with 64 Ω | ZN414Z | | 0.3 | 0.5 | mA |
| headphones | ZN416E | | 4 | 5 | mA |
| Input frequency range | | 0.15 | • | 3.0 | MHz |
| Input resistance | | | 4.0 | | $M\Omega$ |
| Threshold sensitivity (Dependant on Q of coil) | | | 50 | | μV |
| Selectivity Total harmonic distortion | | | 4.0 | | kHz |
| AGC range | | | 3.0 | | % |
| Power gain (ZN414Z) | | | 20 72 | | dB |
| Voltage gain of output stage | ZN416E | | 18 | | dB dB |
| Output voltage into 64Ω load | ZN414Z | | 60 | | mVpp |
| before clipping | ZN416E | | 340 | | mVpp |
| Upper cut-off frequency of output stage | · Y | | | | |
| No capacitor, (ZN416E) With 0.01µF between pin 7 and 0V (ZN416E) | | 20 | | | kHz |
| | | | 10 | | kHz |
| Lower cut-off frequency of output stage | ì | | | | _ |
| 0.47μF between pins 2 and 3 for ZN416E | 3E∫ | | 50 | | kHz |
| Quiescent output voltage | ZN414Z | | 40 | | l mV |
| | ZN416E | | 200 | | mV |
| Operating temperature range | | 0 | | 70 | °C |
| Maximum storage temperature | | -65 | | 125 | °C |
| - · · · | | | | 120 | _ |

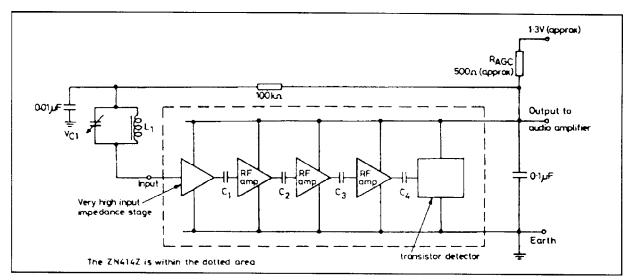


Fig.1 ZN414Z System Diagram

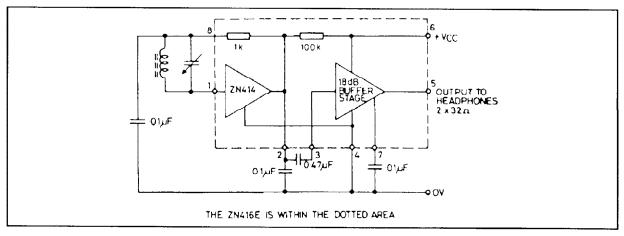


Fig.2 ZN416E System Diagram

ZN414Z CHARACTERISTICS - All measurements performed with 30% modulation, FM = 400Hz

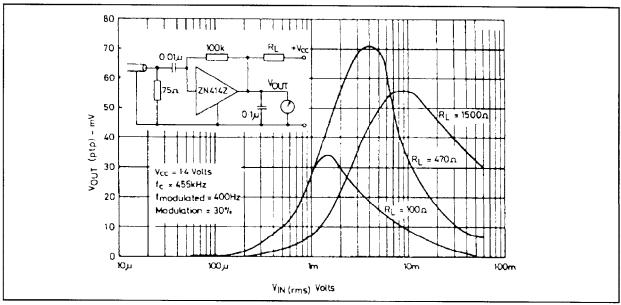


Fig.3 Gain and AGC characteristics

See operating notes for explanation of AGC action.

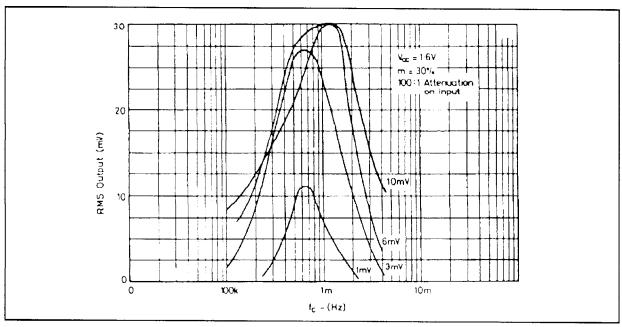


Fig.4 Frequency response of the ZN414Z

Note that this graph represents the chip response, and not the receiver bandwidth.

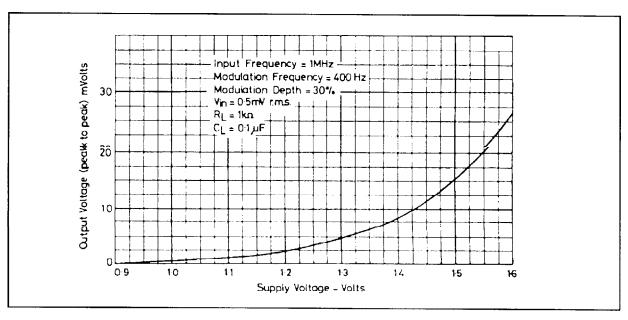


Fig.5 Gain variation with supply volts.

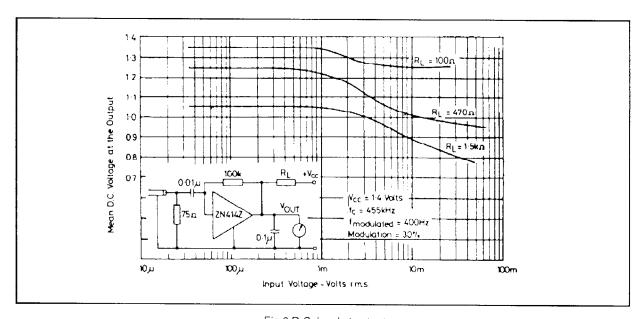


Fig.6 D.C. level at output

LAYOUT REQUIREMENTS

As with any high gain R.F. device, certain basic layout rules must be adhered to if stable and reliable operation is to be obtained. These are listed below:

 The output decoupling capacitor should be soldered as near as possible to the output and earth leads of the ZN414Z. Furthermore, its value together with the AGC resistor (RAGC) should be calculated at =4kHz, i.e.;

C (farads) =
$$\frac{1}{2\pi \cdot \text{Ragc} \cdot 4 \cdot 10^3}$$

- All leads should be kept as short as possible, especially those in close proximity to the ZN414Z.
- 3. The tuning assembly should be some distance from the battery, loudspeaker and their associated leads.
- The 'earthy' side of the tuning capacitor should be connected to the junction of the 100kΩ resistor and the 0.01μF capacitor.

OPERATING NOTES

(a) Selectivity

To obtain good selectivity, essential with any T.R.F. device, the ZN414Z must be fed from an efficient, high 'Q' coil and capacitor tuning network. With suitable components the

selectivity is comparable to superhet designs, except that a very strong signal in proximity to the receiver may swamp the device unless the ferrite rod aerial is rotated to "null-out" the strong signal.

Two otherfactors affect the apparent selectivity of the device. Firstly, the gain of the ZN414Z is voltage sensitive (see previous page) so that, in strong signal areas, less supply voltage will be needed to obtain correct AGC action. Incorrect adjustment of the AGC causes a strong station to occupy a much wider bandwidth than necessary and in extreme cases can cause the RF stages to saturate before the AGC can limit RF gain. This gives the effect of swamping together with reduced AF output. All the above factors have to be considered if optimum performance is to be obtained.

(b) Ferrite aerial size

Because of the gain variation available by altering supply voltage, the size of the ferrite rod is relatively unimportant. However, the ratio of aerial rod length to diameter should ideally be large to give the receiver better directional properties. Successful receivers have been constructed with ferrite rod aerials of 4cm (1.5") and up to 20cm (8").

DRIVE CIRCUITS

Three types of drive circuit are shown, each has been used successfully. The choice is largely an economic one, but circuit 3 is recommended wherever possible, having several advantages over the other circuits. Values for 9V supplies are shown, simple calculations will give values for other supplies.

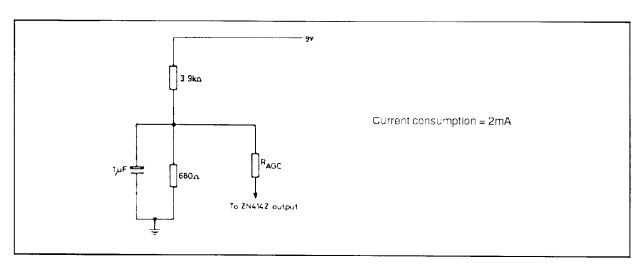


Fig. 7 Resistive Divider (ZN414Z)

Note: Replacing the 680Ω resistor with a 500Ω resistor and a 250Ω preset, sensitivity may be adjusted and will enable optimum reception to be realised under most conditions. reception to be realised under most conditions.

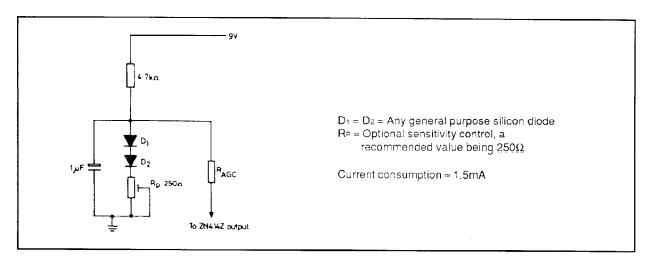


Fig.8 Diode Drive (ZN414Z)

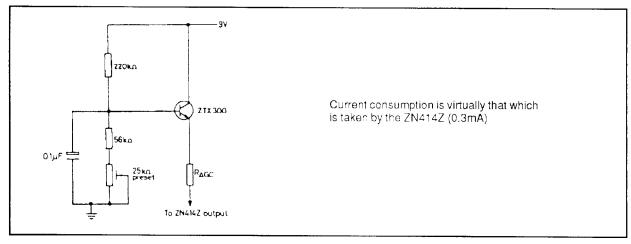


Fig.9 Transistor Drive (ZN414Z and ZN415E)

RECOMMENDED CIRCUITS

(a) Earphone radio

The ZN414Z will drive a sensitive earpiece directly. In this case, an earpiece of equivalent impedance to RAGC is substituted for RAGC in the basic tuner circuit. Unfortunately, the cost of a sensitive earpiece is high, and unless an ultraminiature radio is wanted, it is considerably cheaper to use a low cost crystal earpiece and add a single gain stage. One further advantage of this technique is that provision for a volume control can be made. A suitable circuit is shown below.

 L_1 = 80 turns of 0.3mm dia. enamelled copper wire on a 5cm or 7.5cm long ferrite rod. Do not expect to adhere rigidly to the coil-capacitor details given. Any value of L_1 and C_1 which will give a high 'Q' at the desired frequency may be used. Volume Control: a 250Ω potentiometer in series with a 100Ω fixed resistor substituted for the 270Ω emitter resistor provides an effective volume control.

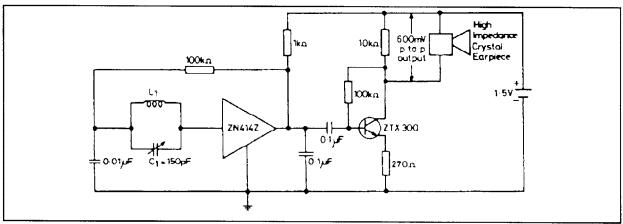


Fig.10 Earphone Radio

(b) Domestic portable receiver

The circuit shown is capable of excellent quality, and its cost relative to conventional designs is much lower

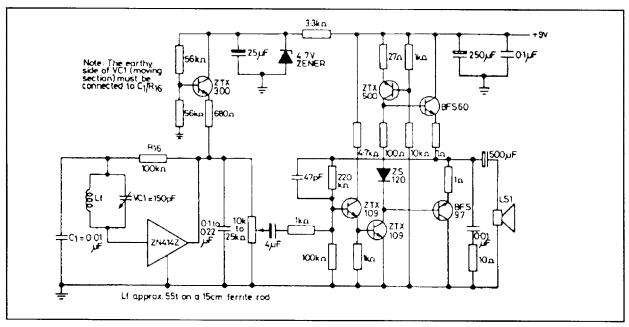


Fig.11 The complete circuit diagram of the Triffid receiver

(b) i

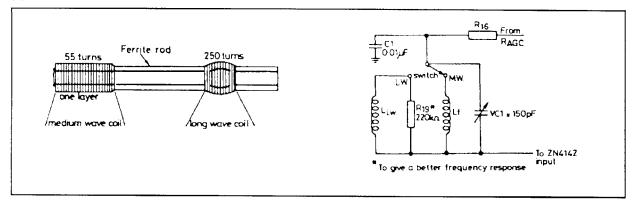


Fig. 12 Coil winding details and waveband selection

(c) Use in model control receiver

The circuit below shows a ZN414Z used as an IF amplifier for a 27MHz superhet receiver.

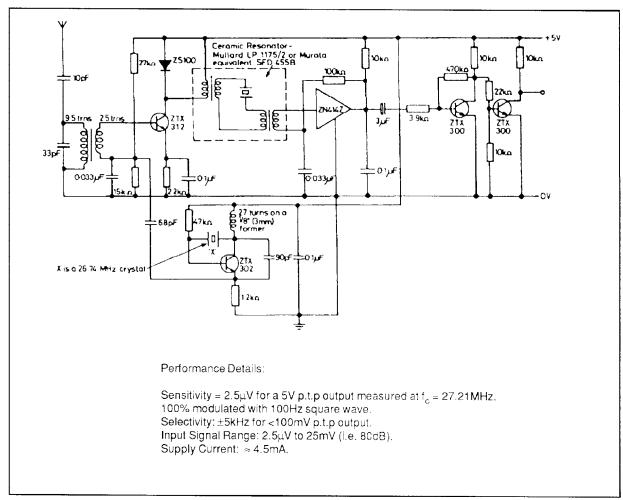


Fig.13

(d) Broadcast band superhet using ZN414Z

The ZN414Z coupled with the modern ceramic resonators offers a very good I.F. amplifier at modest cost, whilst

maintaining simplicity and minimal alignment requirements. A typical circuit is shown below:

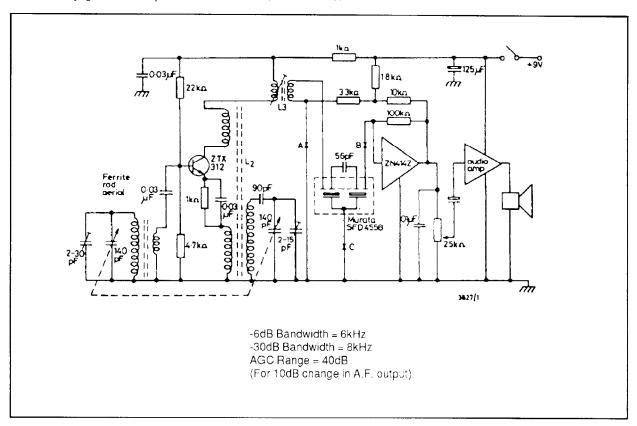


Fig.14

FURTHER APPLICATIONS

The ZN414Z is an extremely versatile device and, in a data sheet, it is not possible to show all its varied applications. A comprehensive applications note on the device is available

which gives full details of various radio receivers, I.F. amplifiers and frequency standards together with comprehensive technical information.



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